GENERAL DESCRIPTION

OB2353 combines a dedicated current mode PWM controller with integrated high voltage power MOSFET. It is optimized for high performance, low standby power consumption, and cost effective off-line flyback converter applications in sub 10W range.

OB2353 operates in fixed PWM frequency, the frequency is set internal and it is accurately trimmed. At no load or light load condition, the frequency is internally adjusted, the IC operates in 'Extended burst mode' to minimize switching loss. Lower standby power consumption and higher conversion efficiency are thus achieved.

VDD low startup current and low operating current contribute to a reliable power-on startup design with OB2353. A large value resistor could thus be used in the startup circuit to minimize the standby power.

The internal slope compensation improves system large signal stability and reduces the possible sub-harmonic oscillation at high PWM duty cycle output. Leading-edge blanking on current sense(CS) input removes the signal glitch due to snubber circuit diode reverse recovery and thus greatly reduces the external component count and system cost in the design.

OB2353 offers complete protection coverage with automatic self-recovery feature including Cycle-by-Cycle current limiting (OCP), over load protection (OLP), VDD over voltage clamp and under voltage lockout (UVLO). Excellent EMI performance is achieved with On-Bright proprietary frequency shuffling technique together with soft switching control at the totem pole gate drive output.

The tone energy at below 20KHZ is minimized in the design and audio noise is eliminated during operation. OB2353 is offered in SOP-8 package.
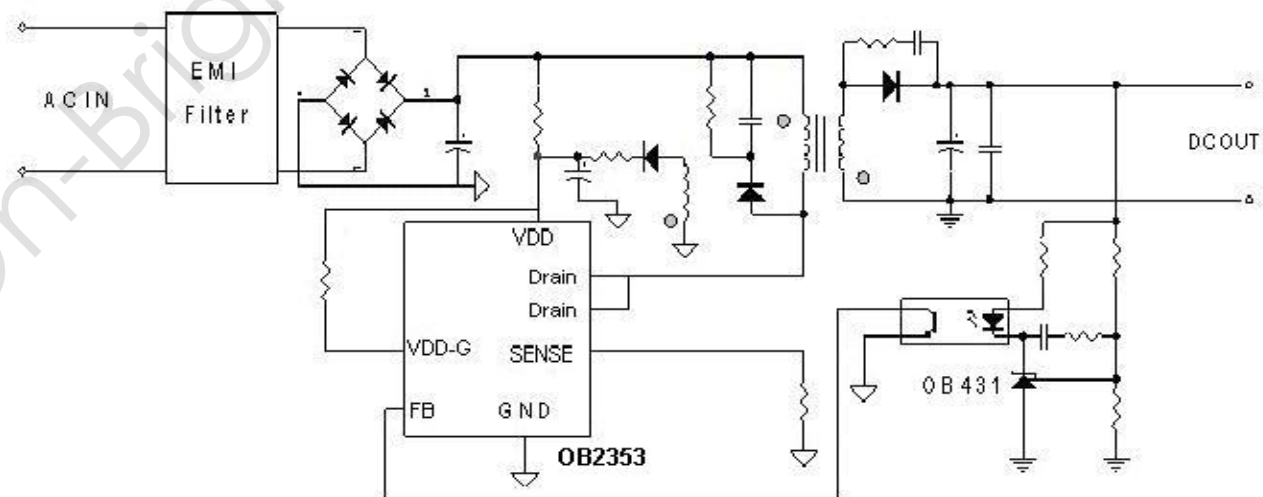
FEATURES

■ Optimized for sub 10W Applications
■ Single Chip 650V Sense FET Power Switch
■ Soft Start
■ Frequency shuffling for EMI
■ Extended Burst Mode Control For Improved Efficiency and Minimum Standby Power Design
■ Audio Noise Free Operation
■ Fixed 50KHZ Switching Frequency
■ Internal Synchronized Slope Compensation
■ Low VDD Startup Current and Low Operating Current
■ Leading Edge Blanking on Current Sense Input
■ Good Protection Coverage With Auto Self-Recovery
 o VDD Over Voltage Clamp and Under Voltage Lockout with Hysteresis (UVLO)
 o On-Bright Proprietary Line Input Compensated Cycle-by-Cycle Over-current Threshold Setting For Constant Output Power Limiting Over Universal Input Voltage Range.
 o Overload Protection (OLP).
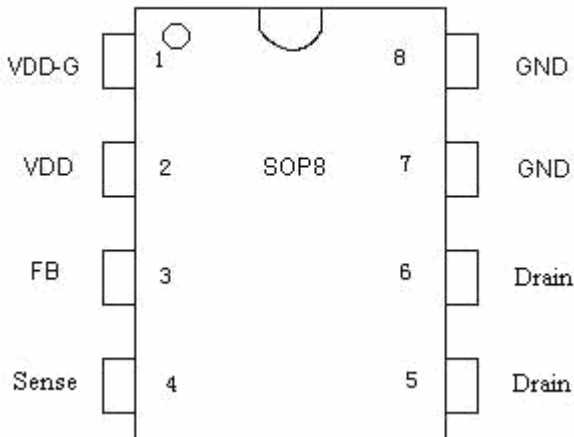 o Over voltage Protection(OVP)

APPLICATIONS

Offline AC/DC flyback converter for
■ Battery Charger
■ PDA power supplies
■ Digital Cameras
■ Power Adaptor
■ Set-Top Box Power Supplies
■ Open-frame SMPS

TYPICAL APPLICATION

GENERAL INFORMATION

Pin Configuration

The OB2353 is offered in SOP8 package, shown as below.

Part Number	Description
OB2353CP	SOP8, Pb-free
OB2353CP/A	SOP8,Pb-free,Tape&Reel

Package Dissipation Rating

	RθJA (°C/W)
SOP8	150

Absolute Maximum Ratings

Parameter	Value
Drain Input Voltage	650V
VDD Input Voltage	30 V
VDD-G Input Voltage	30 V
VDD Input DC Current	10mA
V_{FB} Input Voltage	-0.3 to 7V
V_{SENSE} Input Voltage to Sense Pin	-0.3 to 7V
Min/Max Operating Junction Temperature T_J	-20 to 150 °C
Min/Max Storage Temperature T_{stg}	-55 to 160 °C
Lead Temperature (Soldering, 10secs)	260 °C

Note: Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum-rated conditions for extended periods may affect device reliability.

Ordering Information
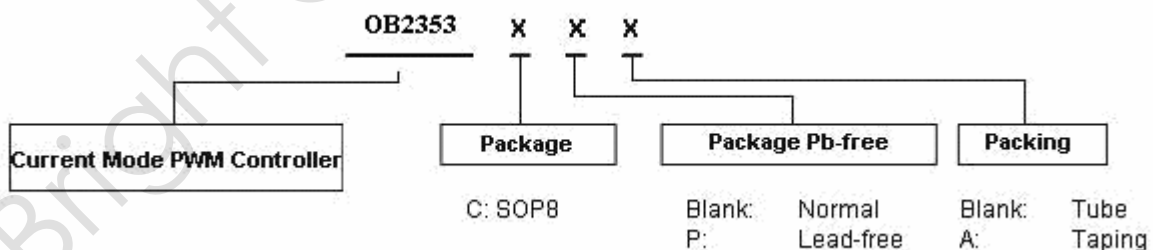
Marking Information

LT YWW

OB2353XP

○   s
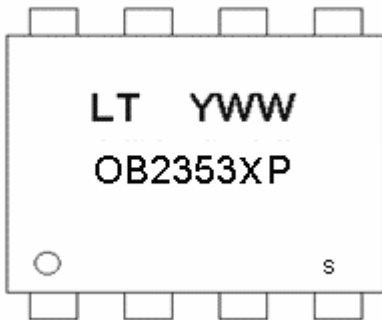
X: C for SOP8;

P: Pb-free Package;

Y: Year Code (0-9)

WW: Week Code (1-52)

s: Optional Internal Code

TERMINAL ASSIGNMENTS

Pin Name	I/O	Description
GND	P	Ground
FB	I	Feedback input pin. The PWM duty cycle is determined by voltage level into this pin and the current-sense signal at Pin 3.
VDD-G	P	Power Supply specifically for Internal Gate Driver
SENSE	I	Current sense input pin.
VDD	P	Chip DC power supply pin.
Drain	P	The Drain pin is designed to connect directly to the primary lead of the transformer and is capable of switching a maximum of 650V

BLOCK DIAGRAM
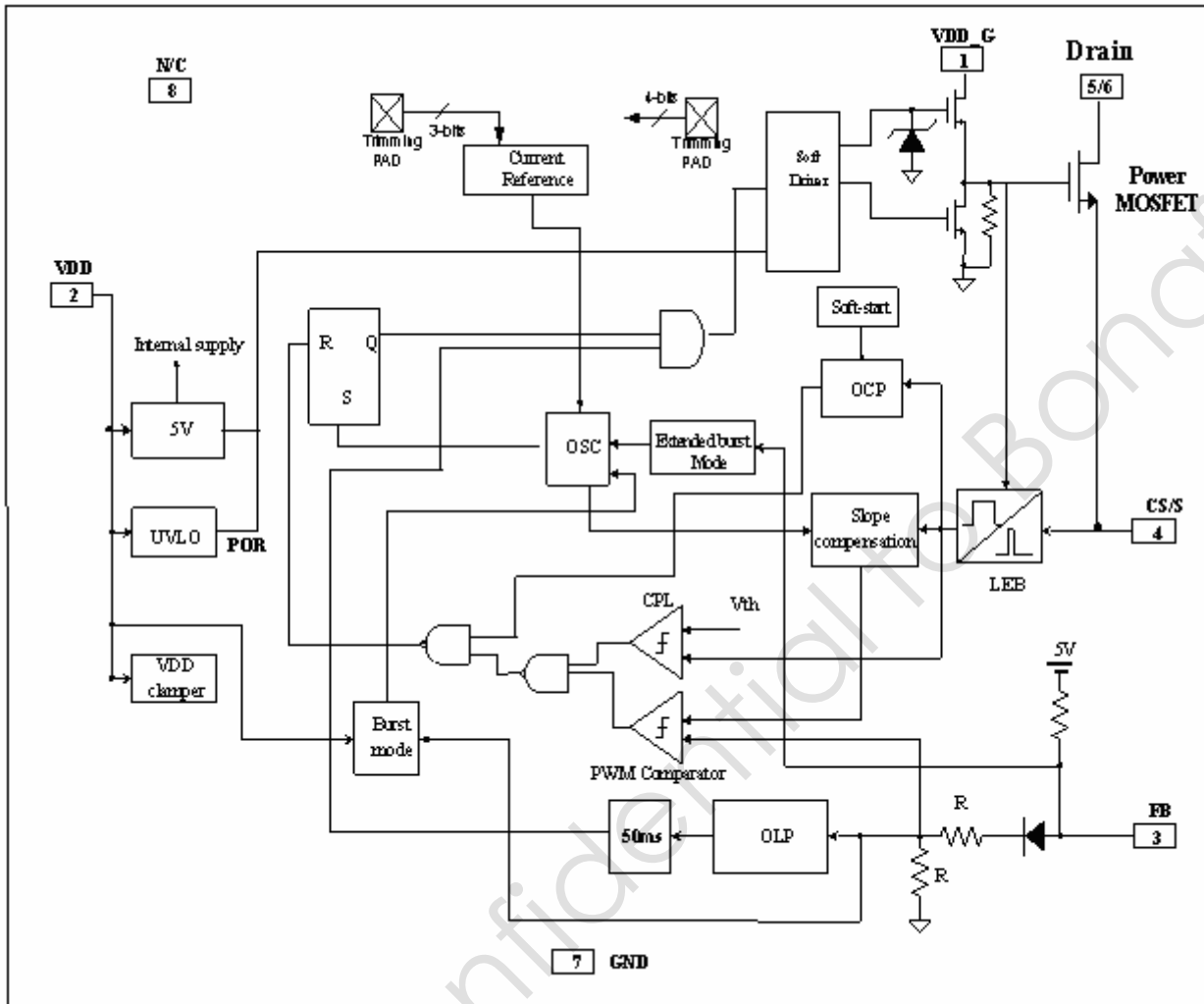
ELECTRICAL CHARACTERISTICS

$(T_A = 25^O C$ if not otherwise noted)

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
Supply Voltage (VDD)						
I_VDD_Startup	VDD Start up Current	VDD =14.5V,Measure Leakage current into VDD		5	20	uA
I_VDD_Operation	Operation Current	VDD=16V, V_{FB}=3V		1.6		mA
UVLO(ON)	VDD Under Voltage Lockout Enter		8.7	9.7	10.7	V
UVLO(OFF)	VDD Under Voltage Lockout Exit (Recovery)		14.6	15.8	17.0	V
OVP(ON)	Over voltage protection voltage	VDD=16V,CS=0V,FB=3V Ramp up VDD until gate clock is off	27.0	28.5	30.0	V
VDD_Clamp	VDD Zener Clamp Voltage	I_{DD} = 10 mA		30.5		V
Feedback Input Section(FB Pin)						
V_{FB}_Open	V_{FB} Open Loop Voltage		5.4	5.7	6.0	V
I_{FB}_Short	FB pin short circuit current	Short FB pin to GND and measure current		1.45		mA
V_{TH}_0D	Zero Duty Cycle FB Threshold Voltage	VDD=16V		0.8		V
V_{TH}_PL	Power Limiting FB Threshold Voltage			3.7		V
T_D_PL	Power limiting Debounce Time			50		mSec
Z_{FB}_IN	Input Impedance			4		Kohm
Current Sense Input(Sense Pin)						
Soft start time				4		ms
T_blanking	Leading edge blanking time			270		ns
Z_{SENSE}_IN	Input Impedance			40		Kohm
T_D_OC	Over Current Detection and Control Delay	From Over Current Occurs till the Gatedrive output start to turn off		120		nSec
V_{TH}_OC	Internal Current Limiting Threshold Voltage	FB=3.3V VDD=16V	0.72	0.77	0.82	V
Oscillator						
F_{OSC}	Normal Oscillation Frequency	VDD = 16V	45	50	55	KHZ
△f_Temp	Frequency Temperature Stability	VDD = 16V		5		%
△f_VDD	Frequency Voltage Stability			5		%

D_max	Maximum duty cycle	FB=3.3V, CS =0V VDD=16V	70	80	90	%
F_Burst	Burst Mode Base Frequency	VDD = 16V		22		KHZ
Mosfet Section						
Vds	Drain-Source Breakdown Voltage		650			V
RDS(on)	Static Drain-Source On-Resistance			12	15	Ω
Idp	Drain current-peak				0.9	A
Frequency Shuffling						
Δf_OSC	Frequency Modulation range /Base frequency		-4		4	%

CHARACTERIZATION PLOTS

(The characteristic graphs are normalized at Ta=25℃)
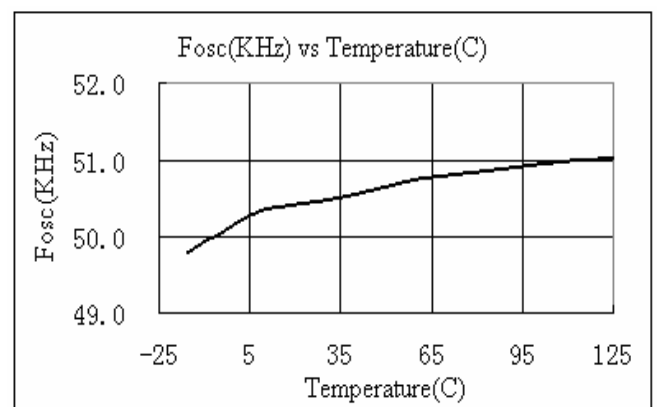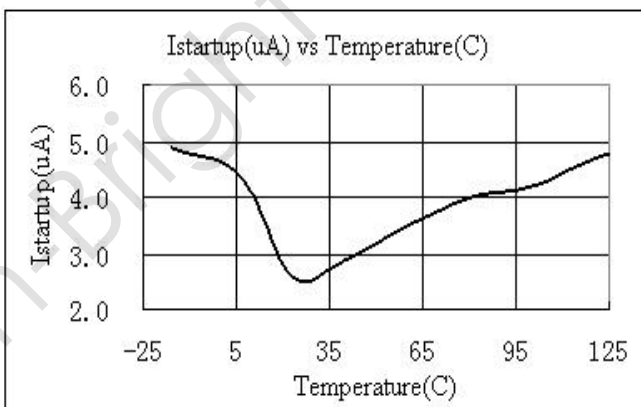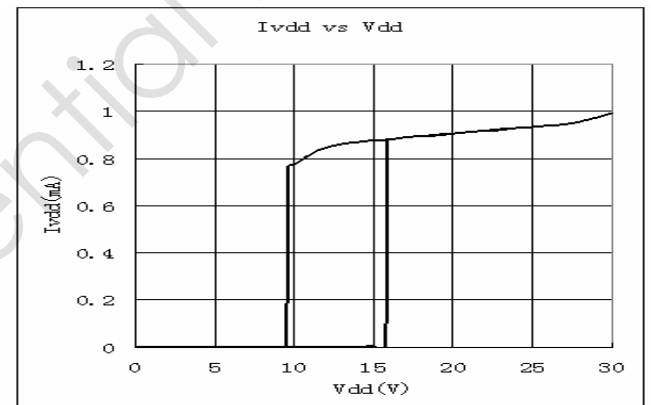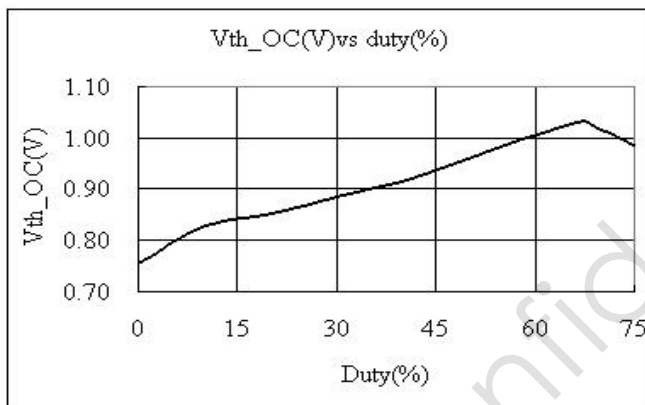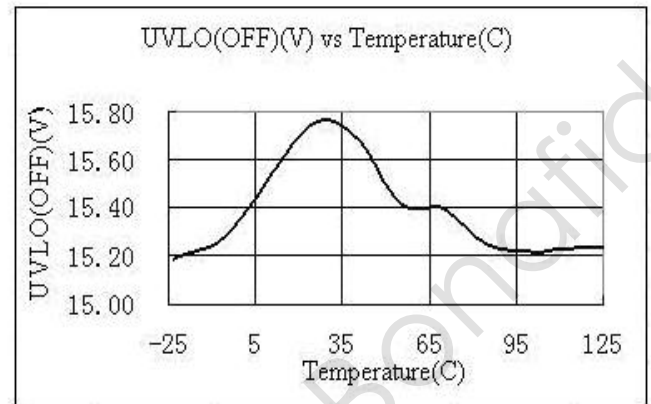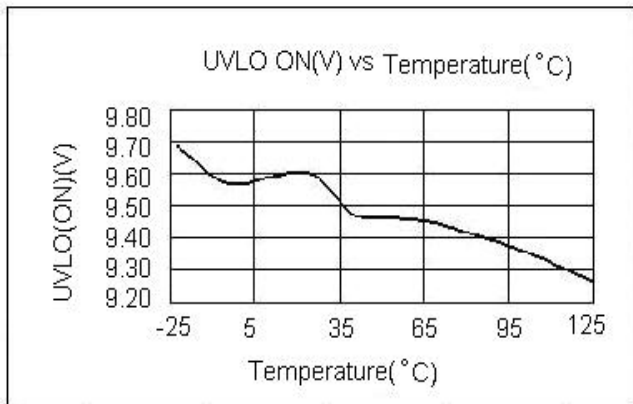
OPERATION DESCRIPTION

The OB2353 is a low power off-line SMPS Switcher optimized for off-line flyback converter applications in sub 10W power range. The 'Extended burst mode' control greatly reduces the standby power consumption and helps the design easily to meet the international power conservation requirements.

- **Startup Current and Start up Control**

Startup current of OB2353 is designed to be very low so that VDD could be charged up above UVLO threshold level and device starts up quickly. A large value startup resistor can therefore be used to minimize the power loss yet reliable startup in application. For AC/DC adaptor with universal input range design, a 2 MΩ, 1/8 W startup resistor could be used together with a VDD capacitor to provide a fast startup and yet low power dissipation design solution.

- **Operating Current**

The Operating current of OB2353 is low at 2mA. Good efficiency is achieved with OB2353 low operating current together with the 'Extended burst mode' control features.

- **Soft Start**

OB2353 features an internal 4ms soft start to soften the electrical stress occurring in the power supply during startup. It is activated during the power on sequence. As soon as VDD reaches UVLO(OFF), the peak current is gradually increased from nearly zero to the maximum level of 0.77V. Every restart up is followed by a soft start.

- **Frequency shuffling for EMI improvement**

The frequency Shuffling (switching frequency modulation) is implemented in OB2353. The oscillation frequency is modulated with a pseudo random source so that the tone energy is spread out. The spread spectrum minimizes the conduction band EMI and therefore eases the system design.

- **Extended Burst Mode Operation**

At light load or zero load condition, most of the power dissipation in a switching mode power supply is from switching loss on the mosfet transistor, the core loss of the transformer and the loss on the snubber circuit. The magnitude of power loss is in proportion to the switching frequency. Lower switching frequency leads to the reduction on the power loss and thus conserves the energy.

The switching frequency is internally adjusted at no load or light load condition. The switch frequency reduces at light/no load condition to improve the conversion efficiency. At light load or no load condition, the FB input drops below burst mode threshold level and device enters Burst Mode control. The Gate drive output switches only when VDD voltage drops below a preset level and FB input is active to output an on state. Otherwise the gate drive remains at off state to minimize the switching loss and reduces the standby power consumption to the greatest extend.

The switching frequency control also eliminates the audio noise at any loading conditions.

- **Oscillator Operation**

The switching frequency of OB2353 is internally fixed at 50KHZ. No external frequency setting components are required for PCB design simplification.

- **Current Sensing and Leading Edge Blanking**

Cycle-by-Cycle current limiting is offered in OB2353 current mode PWM control. The switch current is detected by a sense resistor into the sense pin. An internal leading edge blanking circuit chops off the sensed voltage spike at initial internal power MOSFET on state due to snubber diode reverse recovery and surge gate current of internal power MOSFET so that the external RC filtering on sense input is no longer needed. The current limiting comparator is disabled and cannot turn off the internal power MOSFET during the blanking period. The PWM duty cycle is determined by the current sense input voltage and the FB input voltage.

- **Internal Synchronized Slope Compensation**

Built-in slope compensation circuit adds voltage ramp onto the current sense input voltage for PWM generation. This greatly improves the close loop stability at CCM and prevents the sub-harmonic oscillation and thus reduces the output ripple voltage.

- **Drive**

The internal power MOSFET in OB2353 is driven by a dedicated gate driver for power switch control. Too weak the gate drive strength results in higher

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conduction and switch loss of MOSFET while too strong gate drive results the compromise of EMI

A good tradeoff is achieved through the built-in totem pole gate design with right output strength and dead time control. The low idle loss and good EMI system design is easier to achieve with this dedicated control scheme.

In addition to the gate drive control scheme mentioned, the gate drive strength can also be adjusted externally by a resistor connected between VDD and VDDG, the falling edge of the Drain output can be well controlled. It provides great flexibility for system EMI design.

- **Protection Controls**

Good power supply system reliability is achieved with its rich protection features including Cycle-by-

Cycle current limiting (OCP), Over Load Protection (OLP) and over voltage clamp, Under Voltage Lockout on VDD (UVLO).

With On-Bright Proprietary technology, the OCP is line voltage compensated to achieve constant output power limit over the universal input voltage range.

At overload condition when FB input voltage exceeds power limit threshold value for more than TD_PL, control circuit reacts to shut down the switcher. Switcher restarts when VDD voltage drops below UVLO limit.

VDD is supplied by transformer auxiliary winding output. It is clamped when VDD is higher than 30V. The output of OB2353 is shut down when VDD drops below 9.7V UVLO limit and Switcher enters power on start-up sequence thereafter.

PACKAGE MECHANICAL DATA

8-Pin Plastic SOP
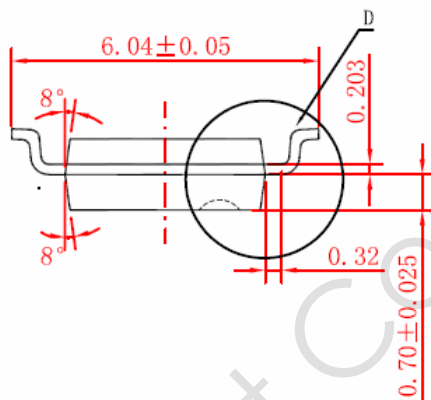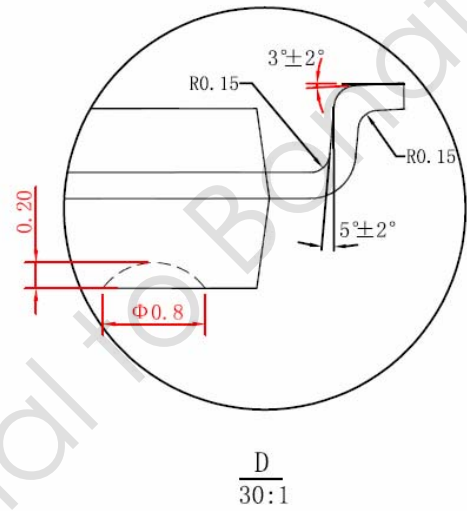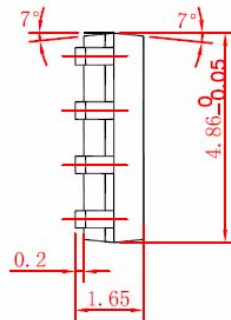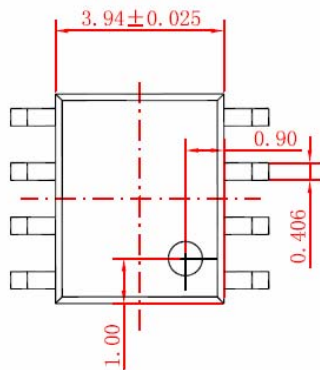
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